

(12) INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(19) World Intellectual Property Organization
International Bureau



(43) International Publication Date
14 June 2001 (14.06.2001)

PCT

(10) International Publication Number
WO 01/43302 A1

(51) International Patent Classification⁷: **H04B 1/707**

(21) International Application Number: **PCT/SE00/02067**

(22) International Filing Date: **24 October 2000 (24.10.2000)**

(25) Filing Language: **English**

(26) Publication Language: **English**

(30) Priority Data:
60/170,080 9 December 1999 (09.12.1999) US
09/522,979 10 March 2000 (10.03.2000) US

(71) Applicant: **TELEFONAKTIEBOLAGET LM ERICSSON (publ) [SE/SE]; S-126 25 Stockholm (SE).**

(72) Inventor: **KARLSSON, Jonas; Marine City B-1305, 391 Shiba-cho, Kanazawa-ku, Yokohama 236-0012 (JP).**

(74) Agent: **ERICSSON RADIO SYSTEMS AB; Ericsson Research, Patent Support Unit, S-164 80 Stockholm (SE).**

(81) Designated States (*national*): AE, AG, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, BZ, CA, CH, CN, CR, CU, CZ, DE, DK, DM, DZ, EE, ES, FI, GB, GD, GE, GH, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MA, MD, MG, MK, MN, MW, MX, MZ, NO, NZ, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TR, TT, TZ, UA, UG, UZ, VN, YU, ZA, ZW.

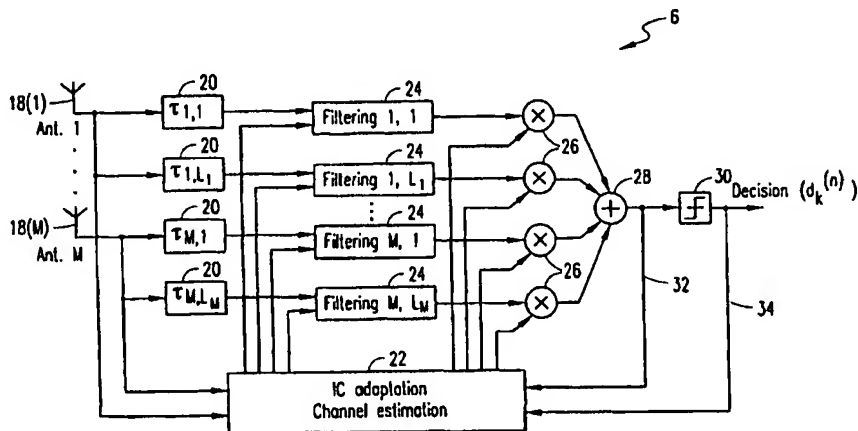
(84) Designated States (*regional*): ARIPO patent (GH, GM, KE, LS, MW, MZ, SD, SL, SZ, TZ, UG, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM), European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GW, ML, MR, NE, SN, TD, TG).

Published:

— *With international search report.*

For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.

(54) Title: **METHOD AND SYSTEM FOR INTERFERENCE CANCELLATION USING MULTIPLE FILTER SETS AND NORMALIZED FILTER ADAPTATION**



(57) Abstract: A method and system for canceling interference in a code division multiple access (CDMA) network. In one embodiment of the invention, a transmitted signal is spread using a specific spreading sequence. The signal includes a plurality of symbols having a specific spreading factor that differs from the length of the spreading sequence. The transmitted signal is received and filtered using a plurality of despreading sequences to recover the transmitted signal, wherein each despreading sequence corresponds to a particular part of the spreading sequence and is used to despread a portion of the received signal that was spread using the corresponding part of the spreading sequence. In addition, each despreading sequence is adapted to cancel interference caused by other signals, wherein such interference is associated with the part of the spreading sequence that corresponds to the despreading sequence. In another embodiment of the invention, a despreading sequence used to decode a received CDMA transmission is adapted to cancel interference, and the adapted despreading sequence is normalized to generate a despreading sequence for use in subsequent filtering.

WO 01/43302 A1

**METHOD AND SYSTEM FOR INTERFERENCE CANCELLATION
USING MULTIPLE FILTER SETS AND NORMALIZED FILTER
ADAPTATION**

CROSS REFERENCE TO RELATED PROVISIONAL APPLICATION

This patent application claims the benefit of priority from, and incorporates by reference the entire disclosure of co-pending U.S. Provisional Patent Application
5 Serial No. 60/170,080 filed on December 9, 1999.

BACKGROUND OF THE INVENTION

Technical Field of the Invention

The present invention relates in general to Code Division Multiple Access
10 (CDMA) communications in cellular radio communications systems, and in particular to interference cancellation in multiple access communications systems.

Description of Related Art

To enable simultaneous communications by multiple users of a cellular system, some type of multiple access technique must be used. In the past, most cellular
15 systems have used either Frequency Division Multiple Access (FDMA), wherein each active call is allocated to a specific frequency, or Time Division Multiple Access (TDMA), wherein each user is allocated a particular time slot and frequency. For the third generation of cellular systems, however, Code Division Multiple Access (CDMA) has been selected as the primary multiple access technique. In particular,
20 Wide-band CDMA (WCDMA) will be used in the majority of the third generation systems.

In CDMA systems, a number of users are allocated to the same frequency. To distinguish among signals for different users, each user is assigned a pseudo-noise spreading sequence, which a transmitter uses to spread the narrow-band information
25 signal to a wide-band signal. To detect a CDMA signal, a conventional receiver despreads the received signal using the same spreading sequence that was used by the transmitter. This despreads the wide-band signal back to a narrow-

- 2 -

band signal. Because the wide-band signal includes information for multiple users, however, the other information signals contained in the wide-band signal tend to interfere with the quality of the despread signal for a particular user. As a result, some form of interference cancellation is typically used to improve performance.

5 One type of interference cancellation uses knowledge of the spreading codes for all of the various users as well as estimated values of the channel characteristics and data symbols for all of the interfering signals. This type of interference cancellation can be implemented using multi-user detection algorithms, such as parallel subtractive interference cancellation. Multi-user detectors are most frequently
10 found in the base stations of cellular systems because all of the required information is generally available in the base stations.

A second type of interference cancellation can be implemented using single-user detectors, which do not have knowledge of the actual spreading sequences used for interfering signals. Only the modulation method and spreading factor are assumed
15 to be known. Single-user detectors are most frequently found in mobile stations because detailed information about the spreading codes or data symbols of the interfering signals is not typically available to the mobile stations. Moreover, single-user detectors are appropriate for mobile stations because they are generally less complex, in terms of the amount of processing resources needed, than multi-user
20 detectors. Because of this reduced complexity, which makes single-user detectors easier to implement, it is sometimes desirable to use single-user detectors in the base stations as well. Interference cancellation in single-user detectors is based on adaptive algorithms that utilize the cyclo-stationary property introduced by the use of short spreading sequences. Accordingly, these adaptive single-user interference cancellation
25 algorithms can effectively cancel interference without knowing the spreading sequences of the interfering signals.

In existing CDMA systems, the short spreading sequence has a periodicity equal to the symbol time. In other words, the length of the spreading sequence is varied to match the spreading factor. Under the WCDMA standard, however, the short
30 spreading sequences, or scrambling codes, have a fixed length of 256 chips regardless of the spreading factor used to transmit information. Thus, the length of the spreading

- 3 -

sequence is only equal to the symbol time when the spreading factor of the data symbols is 256. If the spreading factor is lower (i.e., the data rate is higher), the cyclo-stationary property of the interfering signals is partially destroyed. As a result, existing single-user interference cancellation techniques are not effective in connection with communications using the WCDMA standard.

Moreover, users in the WCDMA environment can be multi-rate users (i.e., the spreading factor for such users can change frequently depending upon the type of application, channel quality, and the like). At any given time, however, different users in a WCDMA system are not necessarily using the same spreading factor. Consequently, interfering signals from other users might have different spreading factors than the user of interest. These problems can further reduce the effectiveness of conventional interference cancellation techniques in the WCDMA environment.

Another problem with many existing adaptive interference cancellation techniques is that they are used to both adapt the receiver to cancel interference and to follow the variations of the radio channel. Because fading on the radio channel can occur very quickly relative to changes in the interfering signals, the adaptation of these interference cancellation techniques often has difficulty following the radio channel variations. Moreover, in addition to handling fading, the adaptive receiver must be able to receive and process signals from multiple antennas and signals propagated through multi-path channels.

There is a need, therefore, for a system and method for canceling interference when the spreading factor used for a particular transmission differs from the length of the short scrambling code for that transmission. The system and method should also be capable of processing transmission signals that are received on multiple antennas and from multiple propagation paths. Furthermore, the system and method would also enable the interference cancellation to adapt to relatively fast variations in the radio channel and to the comparatively slower variations in interference caused by other users' signals. Preferably, the system and method would account for changes in a user's own spreading factor and would provide effective interference cancellation when the spreading factor of interfering users differs from the user's own spreading factor.

- 4 -

SUMMARY OF THE INVENTION

The present invention comprises a method and system for canceling interference in a code division multiple access (CDMA) transmission. In one aspect of the invention, multiple filter sets are used to recover a signal encoded in the transmission, wherein the signal has a spreading factor that differs from the length of a spreading sequence used to encode the signal. In another aspect of the invention, the filter or filters used to recover a signal encoded in the transmission are adapted using a normalized adaptation procedure.

In accordance with one embodiment of the invention, there is provided a method in which a signal encoded in a CDMA transmission is transmitted on a particular channel. The signal includes a plurality of symbols, each having an associated spreading factor, and is spread using a particular spreading sequence. In this embodiment, the length of the spreading sequence differs from the spreading factor associated with at least one of the symbols. Upon receiving the CDMA transmission, it is filtered using a number of despread sequences to recover the encoded signal. Each of the despread sequences corresponds to a particular portion of the spreading sequence. Accordingly, each symbol in the received CDMA transmission is despread using one of the despread sequences that corresponds to the portion of the spreading sequence that was used to spread the symbol. Each despread sequence is further adapted to cancel interference associated with its corresponding portion of the spreading sequence. Generally, such interference is caused by other signals encoded in the CDMA transmission. Preferably, the adaptation step is performed after each filtering operation to generate an updated despread sequence for use in the next filtering operation.

In accordance with another embodiment of the invention, interference cancellation is implemented in a system that includes one or more antennas for receiving a CDMA transmission that includes a signal encoded by a particular spreading sequence. The system includes an interference cancellation adaptation unit for adapting each of a plurality of despread vectors for purposes of canceling interference in the received CDMA transmission. In addition, the system includes one or more filtering units for despread the received CDMA transmission using the

- 5 -

adapted despreading vectors to recover the encoded signal. Each filtering unit comprises a plurality of despreading filters, each of which uses one of the adapted despreading vectors to despread a portion of received CDMA transmission. In particular, the specific adapted despreading vector that is used is the one that
5 corresponds to the part of the spreading sequence used to encode the signal.

In accordance with another embodiment of the invention, a CDMA transmission, in which a signal is encoded using a particular spreading sequence, is transmitted on a specific channel. The CDMA transmission is received and filtered using at least one despreading sequence to recover the signal. Subsequently, the at
10 least one despreading sequence is adapted for purposes of canceling interference caused by other signals in the CDMA transmission. The adapted despreading sequences are further normalized to generate despreading sequences for use in a subsequent filtering step.

As an additional alternative, an error vector can be calculated for each
15 despreading sequence for use in adapting the despreading sequences. Preferably, each such error vector is normalized and then used to calculate the corresponding adapted despreading sequence. Furthermore, the despreading sequences can also be adapted to account for radio channel variations, wherein the adaptation is performed according to an amount of energy and a phase associated with portions of the received CDMA
20 transmission.

In yet another embodiment, there is provided an interference cancellation system having at least one antenna for receiving a CDMA transmission that includes therein a signal encoded by a spreading sequence. The system includes an interference cancellation unit for adapting at least one despreading vector to cancel interference in
25 the received CDMA transmission. The interference cancellation unit also normalizes the adapted despreading sequences to generate updated despreading sequences for use in subsequent filtering. The system further includes one or more filtering units for despreading the received CDMA transmission using the normalized, adapted despreading vectors to recover the encoded signal.

30

- 6 -

BRIEF DESCRIPTION OF THE DRAWINGS

For a more complete understanding of the present invention, reference is made to the following detailed description taken in conjunction with the accompanying drawings wherein:

5 FIGURE 1 is a block diagram of a mobile telecommunications network in which the present invention can be implemented;

 FIGURE 2 is a block diagram of a portion of a WCDMA receiver used for adaptive interference cancellation on received signals;

 FIGURE 3 is a schematic example of a situation where the spreading factor is
10 lower than the length of the spreading sequence;

 FIGURE 4 is a block diagram of an example of a filtering block of FIGURE 2 that includes multiple filter sets for use with different spreading factors;

 FIGURE 5 is a block diagram of an example of a filtering block of FIGURE 2 that includes a single, shared filter set for use with multiple different spreading
15 factors; and

 FIGURE 6 is a schematic example of a situation in which the spreading factor is two times lower than the length of the spreading sequence, but the symbols are despread using a set of four filters.

20 DETAILED DESCRIPTION OF THE INVENTION

Reference is now made to the Drawings wherein like reference characters denote like or similar parts throughout the various Figures. Referring now to FIGURE 1, there is illustrated a block diagram of a mobile telecommunications network 2 in which the present invention can be implemented. The network 2 includes a mobile
25 switching center (MSC) 10 that serves a particular area. The MSC 10 generally controls cellular communications within its serving area and also routes call connections to other MSCs 10 in the network 2 or to external networks such as a public switched telephone network 4. In connection with its control function, the MSC 10 oversees a plurality of base transceiver stations (BTSs) 12 and routes calls
30 involving a particular mobile station (MS) 14 to and from the base transceiver station 12 that serves the cell where that mobile station 14 is currently located.

- 7 -

Communications between each base transceiver station 12 and a plurality of mobile stations 14 are conducted via an air interface 16. Under the WCDMA standard, communications between a base transceiver station 12 and multiple mobile stations 14 are simultaneously conducted on the same bandwidth using spreading codes that are unique for each mobile station 14. As a result, however, the quality of signals between the base transceiver station 12 and a particular mobile station 14 is somewhat diminished by the interference of signals with other mobile stations 14.

Referring now to FIGURE 2, there is illustrated a block diagram of a portion of a WCDMA receiver 6 used for adaptive interference cancellation on received signals. The receiver 6 can be located either in a base transceiver station 12 or in a mobile station 14. In accordance with WCDMA standards, however, short scrambling codes are generally only used in the uplink direction. As a result, adaptive single-user interference cancellation based on the cyclo-stationary properties of the short scrambling codes will most frequently be implemented in the base stations of a WCDMA system.

By using adaptive single-user interference cancellation algorithms, the receiver 6 can cancel interference for each user without interacting with the detection of other users in the system. As a result, interference cancellation in accordance with the present invention can be implemented in a modular way without requiring intercommunication between receivers 6 or between the detection of different users' signals, which simplifies the service, maintenance, and redundancy requirements for the system. Moreover, for interference cancellation in the base stations, these benefits make it easy to gradually upgrade base stations that do not have interference cancellation capabilities to a fully functional interference cancellation base station. In particular, all of the receiver boards in such base stations do not need to be changed at once, and there is no need for additional inter-board communications (i.e., the same backplane can be used). Accordingly, the use of the interference cancellation techniques of the present invention can increase the capacity of base stations, increase the range of base stations, and/or permit the use of lower output power in the mobile terminals.

- 8 -

Returning now to the structure of the receiver 6, the receiver 6 can include one or more antennas 18 that each receive a transmitted signal. When multiple antennas 18 are used, the signals that are received by each antenna 18 can have different propagation delays. In addition, multi-path propagation in the radio environment results in each antenna 18 receiving a plurality of rays. Accordingly, each ray at each antenna 18 should be processed separately. A searcher/tracker unit (not shown) in the receiver 6 identifies and selects the strongest rays received by each antenna 18 and calculates a time delay for each individual ray.

Each time delay from the searcher/tracker unit is used by a corresponding time delay unit 20 to temporally align all of the received rays. The received signals from the antennas 18 are also provided to an interference cancellation (IC) adaptation and channel estimation unit 22, which adapts the despreading vectors to cancel interference and generates channel compensation values to account for variations in the radio channel. To maximize performance, the interference cancellation adaptation is performed relatively slowly because of the slow changes in interference, while the channel compensation values are adapted quickly because of the relatively rapid changes in the radio channel.

The adapted despreading vectors are provided to a corresponding filtering block 24 for each ray. The filtering blocks 24 function to multiply the time-delayed signal rays by the adapted despreading vectors. The outputs of the filtering blocks 24 are further filtered by multiplying them by the channel compensation values at multipliers 26, and the filtered signals for each ray of each antenna are summed at an adder 28. The summed signals are processed by a decision block 30 to generate decision values for the detected signal (e.g., by determining whether each received symbol represents a "0" or a "1"). In addition, the summed signals and the detected signal are provided to the IC adaptation and channel estimation unit 22 by feedback loops 32 and 34 for use in subsequent interference cancellation and radio channel adaptations.

Mathematically, the transmitted radio signal received by antenna m can be represented by:

- 9 -

$$r_m(t) = \sum_{k=1}^K \sum_{n=0}^{N_k-1} \sum_{l=1}^{L_k} b_k^{(n)} c_{k,l,m}^{(n)} s_k^{(n)}(t - nT_k - \tau_{k,l,m}) + n_m(t), \quad (1)$$

where K is the number of users, N_k is the number of symbols transmitted by user k , L_k is the number of rays received from user k , $b_k^{(n)}$ are the transmitted symbols, $c_{k,l,m}^{(n)}$ is the complex channel coefficient, $s_k^{(n)}(t)$ is the spreading sequence for user k 's n th symbol (wherein the duration of the spreading sequence is T_k), T_k is the symbol time of user k , $\tau_{k,l,m}$ is the time delay of the rays, and $n_m(t)$ is the noise received by the m th antenna 18 as a function of time.

The time delay blocks 20 generate a vector $\mathbf{r}_{k,l,m}^{(n)}$ that contains the time aligned samples corresponding to user k 's n th symbol, which consists of SF_k chips. In the case of chip-rate sampling (i.e., where the received signal is sampled once for each chip), the i th element (wherein SF_k is the spreading factor for user k and $i = 0, \dots, (SF_k - 1)$) in the vector $\mathbf{r}_{k,l,m}^{(n)}$ is $r_m(nT_k + \tau_{k,l,m} + T_c i)$, where T_c is the chip period. Thus, the symbol time for user k is $T_k = SF_k * T_c$. The IC adaptation and channel estimation unit 22 provides filtering vectors $\mathbf{w}_{k,l,m}^{(n)}$ to the filtering blocks 24 and channel estimates $\hat{c}_{k,l,m}^{(n)}$ to the multipliers 26 for user k . The receiver 6 then generates a detected signal $d_k^{(n)}$ given by:

$$d_k^{(n)} = \sum_{m=1}^M \sum_{l=1}^{L_k} \hat{c}_{k,l,m}^{(n)H} \mathbf{w}_{k,l,m}^{(n)H} \mathbf{r}_{k,l,m}^{(n)} \quad (2)$$

where $(\bullet)^H$ is the complex conjugate matrix transpose operation (i.e., $\hat{c}_{k,l,m}^{(n)H}$ is the complex conjugate matrix transpose of $\hat{c}_{k,l,m}^{(n)}$ and $\mathbf{w}_{k,l,m}^{(n)H}$ is the complex conjugate matrix transpose of $\mathbf{w}_{k,l,m}^{(n)}$). In the case of BPSK modulation, the final decision on the detected signal has the sign of the real part of the decision variable $d_k^{(n)}$.

To calculate the channel estimates $\hat{c}_{k,l,m}^{(n)}$ many different methods are known in the art. In a preferred embodiment, the channel estimates are adapted to account for radio channel variations in accordance with an amplitude and a phase associated with portions of the received radio transmission. In addition, many different methods are available to calculate the filtering vectors $\mathbf{w}_{k,l,m}^{(n)}$. Preferably, however, the new

- 10 -

filtering vector is calculated based on the previous filtering vector and other information such that:

$$w_{k,l,m}^{(n)} = f(w_{k,l,m}^{(n-1)}). \quad (3)$$

5

This is possible because, for each user k , the spreading sequence $s_k^{(n)}(t)$ is the same for all symbols n , assuming the short spreading sequence has a periodicity equal to the symbol time.

In the case of a WCDMA receiver 6, however, this assumption does not always hold. As mentioned above, the short spreading sequences have a fixed length of 256 chips regardless of the spreading factor used to transmit information. Even though the symbol-wise cyclo-stationary property will be partly destroyed in cases where the spreading factor differs from the length of the spreading sequence, other periods of cyclo-stationarity will still exist. Referring now to FIGURE 3, there is illustrated a schematic example of a situation where the spreading factor 42 is lower than the length of the spreading sequence 40. In particular, the spreading factor in this case is 64 (i.e., the spreading factor is four times lower than the length of the spreading sequence). As a result, every fourth symbol 44 is spread using the same portion of the spreading sequence.

In accordance with the present invention, four independent adaptation filters are used to process the symbols 44. In other words, the received symbols 44 can be grouped into four different sets (i.e., in Set 1, $n = 0, 4, 8, 12, 16, \dots$; in Set 2, $n = 1, 5, 9, 13, 17, \dots$; in Set 3, $n = 2, 6, 10, 14, 18, \dots$; and in Set 4, $n = 3, 7, 11, 15, 19, \dots$; where n is the time index for the symbols). Each set then has a corresponding adaptation filter that is used for canceling interference on the symbols in that set.

Although the example depicted in FIGURE 3 uses a spreading factor that is four times lower than the length of the spreading sequence, other spreading factors can also be used. Accordingly, if the length of the spreading sequence 40 is denoted by SFC (in the WCDMA standard, $SFC = 256$) and if the spreading factor for user k is SFD_k , the filtering vector at time n (i.e., $w_{k,l,m}^{(n)}$) is calculated based on the filtering vector at time $n - (SFC / SFD_k)$ (i.e., $w_{k,l,m}^{(n - (SFC / SFD_k))}$) and other information, such that:

30

- 11 -

$$w_{k,l,m}^{(n)} = f(w_{k,l,m}^{(n-(SFC/SFD_k))}). \quad (4)$$

Generally, SFC / SFD_k is assumed to be an integer number. Thus, the WCDMA receiver 6 will have SFC / SFD_k number of filtering or despreading vectors that are updated independently to account for the situation in which the spreading factor 42 differs from the length of the spreading sequence 40.

However, simply having one set of filters for such a situation is not always sufficient. In many cases, users can change spreading factors frequently. For example, the spreading factor might be different for different services or might change as a result of variations in the radio channel quality. In accordance with another aspect of the invention, the number of adaptation filters used in the receiver 6 change when the user changes spreading factors. Accordingly, a separate set of filters is maintained for each different spreading factor that is used by the user. For example, a set of four adaptation filters is maintained for use when $SFC / SFD_k = 4$, while a set of eight adaptation filters is maintained for use when $SFC / SFD_k = 8$.

Referring now to FIGURE 4, there is illustrated a block diagram of an example of a filtering block 24 of FIGURE 2 that includes multiple filter sets for use with different spreading factors. A first filter set 50 is designed for use when $SFC / SFD_k = 4$. The first filter set 50 includes four adaptation filters 54(1)-54(4) that correspond to four different parts of the spreading sequence. Each adaptation filter 54 receives from the IC adaptation unit 22 an adaptation vector that is used for despreading the set of symbols corresponding to that portion of the spreading sequence. In addition, the adaptation filters 54 receive the appropriate time-delayed signal ray from a time delay unit 20, which is then filtered using the received adaptation vectors. The outputs of the adaptation filters 54 are then sequentially combined by an adaptation filter multiplexer 56, and the combined output is sent via a filter set multiplexer 61 to a channel estimation multiplier 26 for that ray.

Similarly, a second filter set 52 is designed for use when $SFC / SFD_k = 8$. The second filter set 52 includes eight adaptation filters 58(1)-58(8) that correspond to eight different parts of the spreading sequence. Each adaptation filter 58 receives from the IC adaptation unit 22 an adaptation vector that is used for despreading the set of

- 12 -

symbols corresponding to that portion of the spreading sequence. In addition, the adaptation filters 58 receive the appropriate time-delayed signal ray from a time delay unit 20, which is then filtered using the received adaptation vectors. The outputs of the adaptation filters 58 are then sequentially combined by an adaptation filter multiplexer 60, and the combined output is sent via the filter set multiplexer 61 to a channel estimation multiplier 26 for that ray. Although not depicted in the Figure, the filtering block 24 can include other sets of filters for supporting interference cancellation in connection with other spreading factors. Accordingly, separate filter sets can be maintained for each of a particular user's different spreading factors. In any event, the filter set multiplexer 61 selects, at any given time, the signal from only one of the adaptation filter multiplexers 56 and 60 depending on the current value of SFC / SFD_k . Thus, the filter set multiplexer 61 selects the signal from the adaptation filter multiplexer 56 for the first filter set 50 when $SFC / SFD_k = 4$ and the signal from the adaptation filter multiplexer 60 for the second filter set 52 when $SFC / SFD_k = 8$.

In another embodiment of the invention, if the number of spreading factors that are used by a particular user is large, the number of filter sets can be reduced by sharing filter sets among spreading factors that are close to each other. This reduction can help reduce the complexity of the interference cancellation processing and can help avoid a situation in which the convergence time becomes too long. For example, a set of four adaptation filters, each corresponding to one quarter of the spreading sequence, might be used both when $SFC / SFD_k = 2$ and when $SFC / SFD_k = 4$, while a set of sixteen adaptation filters, each corresponding to one sixteenth of the spreading sequence, might be used both when $SFC / SFD_k = 8$ and when $SFC / SFD_k = 16$. As a result, only two filter sets would be needed in this example instead of four.

Referring now to FIGURE 5, there is illustrated a block diagram of an example of a filtering block 24 of FIGURE 2 that includes a single, shared filter set for use with multiple different spreading factors. The filtering block 24 includes sixteen adaptation filters 70-85, each corresponding to one sixteenth of the spreading sequence. These filters, for example, can be used both when $SFC / SFD_k = 8$ and when $SFC / SFD_k = 16$. Each adaptation filter 70-85 receives from the IC adaptation unit 22 an adaptation vector that is used for despreading the set of symbols corresponding to that portion of

- 13 -

the spreading sequence. In addition, the adaptation filters 70-85 receive the appropriate time-delayed signal ray from a time delay unit 20 via a de-multiplexer 68. The de-multiplexer 68 serves to separate the ray into sequential segments that each correspond to a particular part of the spreading sequence. The de-multiplexer 68 then
 5 sends each segment to the corresponding adaptation filter 70-85 for that part of the spreading sequence. Each adaptation filter 70-85 filters the received time-delayed signal ray using the received adaptation vectors for the corresponding part of the spreading sequence. Thus, each of the adaptation filters 70-85 filters one symbol when $SFC / SFD_k = 16$ and half a symbol when $SFC / SFD_k = 8$. The outputs of the
 10 adaptation filters 70-85 are then sequentially combined by a multiplexer 87, and the combined output is sent to a channel estimation multiplier 26 for the ray.

Although the adaptation filters 70-85 are described as being used when $SFC / SFD_k = 8$ and when $SFC / SFD_k = 16$, the same adaptation filters 70-85 can also be used for other values of SFC / SFD_k , such as $SFC / SFD_k = 1, 2$, or 4. The
 15 performance of the adaptative filtering tends to deteriorate, however, as the value of SFC / SFD_k becomes much smaller than the number of filters in the filter set. Accordingly, when filter sets are shared among different spreading factors, each particular filter set is preferably only shared with spreading factors that are close to the optimum spreading factor for that filter set (e.g., wherein the optimum spreading factor
 20 is $SFC /$ the number of adaptation filters; thus, for a filter set that includes sixteen adaptation filters, the optimum spreading factor is $256 / 16 = 16$).

Referring now to FIGURE 6, there is illustrated a schematic example of a situation where the spreading factor 42 is two times lower than the length of the spreading sequence 40, but the symbols are despread using a set of four filters. In this
 25 case, each symbol is divided in half, and each half is despread by a different filter 54 (see FIGURE 4). Thus, a first half 46(1) of the first symbol 42 is sent to a first filter 54(1); a second half 46(2) of the first symbol 42 is sent to a second filter 54(2); a first half 46(3) of the second symbol 42 is sent to a third filter 54(3); and a second half 46(4) of the second symbol 42 is sent to a fourth filter 54(4). As a result, each filter
 30 54 corresponds to a distinct part of the spreading sequence.

- 14 -

In addition to the possibility of frequent changes in a user's own spreading factor, it is also likely that at least some of the interfering users in the system will be using a different spreading factor than that of the user of interest. Although the receiver 6 of the present invention does not include any separate functionality for handling interfering users' varying spreading factors, the interference cancellation procedures of the present invention should effectively cancel interference even when the spreading factors of the other users differ from the one being used by the user of interest.

In another aspect of the invention, the convergence properties of interference cancellation procedures can be improved. As discussed in connection with FIGURE 2, the channel compensation procedures are separated from the interference cancellation filtering. While this separation helps the receiver 6 to better handle rapid fading on the radio channel, it is not always sufficient to obtain a desirable level of convergence for fading radio channels. In accordance with another embodiment of the invention, however, the level of convergence can be improved by placing limitations on the IC adaptation, such as by normalizing the vectors that are used to update the IC filter coefficients.

Typically, the current filter coefficients (i.e., at time $t + 1$) can be determined as a function of a preceding set of filter coefficients (i.e., at time t) such that:

$$w_k(t+1) = w_k(t) + \mu f() \quad (5)$$

where the function $f()$ is an error function that depends upon the selected method of adaptation and μ is the step size for the selected method of adaptation. In accordance with Griffiths' algorithm, for example, the error function (i.e., $f() = e_k(t)$) is given by:

$$e_k(t) = s_k(t) - r(t)\tilde{y}_k^H(t) \quad (6)$$

where

$$\tilde{y}_k(t) = \frac{1}{h_k^H h_k} w_k^H(t) r(t), \quad (7)$$

- 15 -

and h_k is a complex value that represents a channel estimate for user k . This type of algorithm has problems with convergence on fading radio channels.

By placing some limitations on the updating method, however, the convergence properties of the filter coefficient updating can be improved. In a preferred embodiment, both the error function and the adaptation filter coefficients are normalized. Thus, updating of the filter coefficients can be performed according to the following algorithms:

$$w_{temp} = w_k(t) + \mu \frac{f()}{|f()|}, \text{ and} \quad (8)$$

$$w_k(t+1) = \frac{w_{temp}}{|w_{temp}|}. \quad (9)$$

In the case of a multi-path radio channel and/or multiple receiving antennas 18, the normalization of the error function can be a function of the channel estimates of the different rays. Accordingly, if there are M antennas each receiving L rays, the filter coefficients for antenna m and ray l can be updated as follows:

$$w_{temp,m,l} = w_{m,l}(t) + \mu \frac{|h_{m,l}|^2}{\sum_{x=1}^M \sum_{y=1}^L |h_{x,y}|^2} \frac{f_{m,l}()}{|f_{m,l}()|}, \text{ and} \quad (10)$$

$$w(t+1) = \frac{w_{temp,m,l}}{|w_{temp,m,l}|}, \quad (11)$$

where $h_{m,l}$ is the channel estimate for antenna m and ray l . As a result, the actual step size is adjusted to the energy of that particular ray. Alternatively, the actual step size can be adjusted in other ways, such as according to the amplitude of the particular ray

- 16 -

or any other function of the energy of the ray. Similarly, in the case of a single ray, the normalization of the error function can be a function of an amount of energy associated with the ray.

5 A number of different adaptation algorithms can be used in accordance with the normalization and/or multiple filter set aspects of the present invention, including the least mean squares (LMS) algorithm, the recursive least squares (RLS) algorithm, and the normalized recursive least squares (NRLS) algorithm. In one preferred embodiment, however, a normalized Griffiths' algorithm is used for updating the despreading vectors. Thus, the error function is given by:

10

$$e_k(t) = \frac{s_k - r(t)\tilde{y}_k^H(t)}{|s_k - r(t)\tilde{y}_k^H(t)|}, \quad (12)$$

15 and the despreading vector is adapted according to:

$$w_k(t+1) = \frac{w_k(t) + \mu e_k(t)}{|w_k(t) + \mu e_k(t)|}, \quad (13)$$

20 where

$$\tilde{y}_k(t) = \frac{1}{h_k^H h_k} w_k^H(t) r(t), \quad (14)$$

and, in the case of BPSK modulation,

25

$$\hat{d}_k(t) = \text{sign}(\text{real}(h_k^H \tilde{y}_k(t))). \quad (15)$$

In the receiver 6 of FIGURE 2, $\tilde{y}_k(t)$ represents the output of the adder 28, while $w_k(t+1)$ represents the adapted despreading vector generated by the IC adaptation unit 22. In connection with the normalized Griffiths' algorithm, a step size
30 of approximately $\mu = 0.01$ is preferred. Ideally, the adaptation of the despreading

- 17 -

vectors for all of the rays of all of the antennas is performed jointly. The adaptation is much simpler to perform, however, if it is done separately for each ray.

As will be appreciated by those of ordinary skill in the art, the normalized adaptation algorithms can also be used in connection with one or more sets of filters (i.e., when the spreading factor differs from the length of the spreading sequence).
 5 Accordingly, if each value of t corresponds to a transmitted symbol, then updating of the adaptation filters could be accomplished using:

$$10 \quad w_k(t) = \frac{w_k(t - SFC / SFD_k) + \mu \frac{f(t)}{|f(t)|}}{\left| w_k(t - SFC / SFD_k) + \mu \frac{f(t)}{|f(t)|} \right|} \quad (16)$$

15

Although a preferred embodiment of the method and apparatus of the present invention has been illustrated in the accompanying Drawings and described in the foregoing Detailed Description, it is understood that the invention is not limited to the embodiment disclosed, but is capable of numerous rearrangements, modifications, and
 20 substitutions without departing from the spirit of the invention as set forth and defined by the following claims.

- 18 -

WHAT IS CLAIMED IS:

1. A method for canceling interference in a code division multiple access (CDMA) communication system to recover a signal encoded in a CDMA transmission, comprising the steps of:
 - 5 transmitting on a particular channel a CDMA transmission having a specific signal encoded therein, said signal including a plurality of symbols, each symbol having an associated spreading factor, and said signal spread using a particular spreading sequence, wherein the length of the spreading sequence differs from at least one of the associated spreading factors and wherein said CDMA transmission includes
10 at least one other signal encoded therein;
receiving the CDMA transmission having the specific signal encoded therein;
filtering the CDMA transmission using a plurality of despreading sequences to recover the specific signal, wherein each of the plurality of despreading
15 sequences corresponds to a particular portion of the spreading sequence, each symbol despread using at least one of the plurality of despreading sequences, said at least one of the plurality of despreading sequences selected for each symbol according to the portion of the spreading sequence used to spread that symbol; and
adapting each despreading sequence to cancel interference associated
20 with the portion of the spreading sequence that corresponds to that despreading sequence, said interference caused by the at least one other signal.
2. The method of claim 1, wherein a first subset of the plurality of despreading sequences is associated with a first spreading factor and a second subset
25 of the plurality of despreading sequences is associated with a second spreading factor, wherein said first spreading factor differs from said second spreading factor.
3. The method of claim 2, wherein the first subset of the despreading sequences consists of a number of despreading sequences equal to a length of the
30 particular spreading sequence divided by the first spreading factor and the second subset of the despreading sequences consists of a number of despreading sequences

- 19 -

equal to a length of the particular spreading sequence divided by the second spreading factor.

4. The method of claim 2, wherein the first subset of the despreading sequences is further associated with a third spreading factor, said third spreading factor different from said first spreading factor, each symbol associated with the third spreading factor despread using the despreading sequences in the first subset of despreading sequences.

5. The method of claim 1, wherein the particular spreading sequence comprises a repeating short spreading sequence.

6. The method of claim 5, wherein the step of adapting the despreading sequences further comprises updating each despreading sequence based on a calculated error associated with a prior filtering operation involving that despreading sequence.

7. The method of claim 6, wherein said prior filtering operation comprises a despreading of a prior symbol in the specific signal using that despreading sequence.

8. The method of claim 1, further comprising the steps of:
determining at least one channel compensation value associated with the CDMA transmission; and
modifying at least one despread symbol using the at least one channel compensation value.

9. The method of claim 1, wherein the step of receiving comprises receiving a plurality of rays of the CDMA transmission and the step of filtering comprises separately filtering each of the plurality of rays.

10. The method of claim 1, wherein the step of adapting includes placing limits on the adaptation to obtain improved convergence of the adaptation.

- 20 -

11. The method of claim 10, wherein the step of placing limits on the adaptation comprises normalizing the despread sequences.

5 12. The method of claim 11, wherein the step of placing limits on the adaptation further comprises normalizing an error function for each despread sequence, said error function associated with a prior filtering operation involving that despread sequence, and the step of adapting the despread sequences further comprises updating each despread sequence based on the error function.

10 13. The method of claim 12, wherein the step of receiving comprises receiving a plurality of rays of the CDMA transmission, the step of filtering comprises separately filtering each of the plurality of rays, the step of adapting comprises separately adapting the despread sequence for each ray, and the normalization of each error function includes adjusting a step size for said updating based on the energy
15 of each ray.

14. The method of claim 1, wherein said step of adapting the despread sequences comprises updating the despread sequences in accordance with Griffiths' algorithm.

20

15. An interference cancellation system for a CDMA receiver, comprising:
at least one antenna for receiving a CDMA transmission, said CDMA transmission including a signal encoded by a particular spreading sequence;
an interference cancellation adaptation unit for adapting a plurality of
25 despread vectors to cancel interference in the received CDMA transmission, each despread vector corresponding to a specific part of the spreading sequence;
at least one filtering unit for despread the received CDMA transmission using the adapted despread vectors to recover the encoded signal, each filtering unit comprising a set of despread filters, each said despread filter using
30 one of the adapted despread vectors to despread a portion of the received CDMA

- 21 -

transmission encoded by the specific part of the spreading sequence that corresponds to the adapted despreading vector.

16. The system of claim 15, wherein the plurality of despreading vectors
5 consists of a number of despreading vectors equal to a length of the spreading sequence divided by the length of the despreading vectors, wherein all of the despreading vectors are of the same length.

17. The system of claim 15, wherein the interference cancellation
10 adaptation unit adapts each of the plurality of despreading vectors based on an error associated with a previous despreading operation performed using the despreading vector.

18. The system of claim 15, wherein each filtering unit comprises a
15 plurality of sets of despreading filters, each set of despreading filters including a plurality of despreading vectors and associated with at least one spreading factor.

19. The system of claim 15, wherein the at least one antenna receives the
CDMA transmission in a plurality of rays, the system further comprising:
20 a time delay unit for each of the plurality of rays, each said time delay unit for temporally aligning the rays; and
a plurality of filtering units, each filtering unit associated with one of the rays for despreading the received CDMA transmission in that ray.

20. A method for canceling interference in a code division multiple access
25 (CDMA) communication system to recover a signal encoded in a CDMA transmission, comprising the steps of:

transmitting on a particular channel a CDMA transmission having a
specific signal encoded therein, said signal encoded using a particular spreading
30 sequence, wherein said CDMA transmission includes at least one other signal encoded therein;

- 22 -

receiving the CDMA transmission having the specific signal encoded therein;

filtering the CDMA transmission using at least one despreading sequence to recover the specific signal;

5 adapting the at least one despreading sequence for purposes of canceling interference caused by the at least one other signal; and

normalizing the at least one adapted despreading sequence to generate at least one despreading sequence for use in a subsequent filtering step.

10 21. The method of claim 20, further comprising the step of adapting the at least one despreading sequence to account for radio channel variations, said step of adapting to account for radio channel variations performed according to an amplitude and a phase associated with portions of the received CDMA transmission.

15 22. The method of claim 20, wherein the step of adapting to cancel interference further comprises calculating an error vector for each despreading sequence and normalizing each error vector, each normalized error vector used to calculate a corresponding adapted despreading sequence.

20 23. The method of claim 22, wherein the steps of adapting and normalizing comprise using a normalized Griffiths' algorithm to generate the at least one despreading sequence for use in a subsequent filtering step.

25 24. The method of claim 22, wherein the step of filtering the CDMA transmission comprises using a plurality of despreading sequences to recover the specific signal, each despreading sequence corresponding to a specific part of the spreading sequence, wherein each despreading sequence is used to despread at least one portion of the CDMA transmission encoded using the corresponding specific part of the spreading sequence.

30

- 23 -

25. The method of claim 24, wherein the plurality of despread sequences comprise a number of despread sequences equal to a length of the spreading sequence divided by a length of the despread sequences.

- 5 26. An interference cancellation system for a CDMA receiver, comprising:
 at least one antenna for receiving a CDMA transmission, said CDMA
transmission including a signal encoded by a particular spreading sequence;
 an interference cancellation adaptation unit for adapting at least one
despread vector to cancel interference in the received CDMA transmission, wherein
10 the interference cancellation unit further normalizes the at least one adapted
despread sequence to generate at least one despread sequence for use in
subsequent filtering; and
 at least one filtering unit for despread the received CDMA
transmission using the at least one normalized, adapted despread vector to recover
15 the encoded signal.

27. The system of claim 26, wherein each of a plurality of despread
vectors corresponds to a specific part of the spreading sequence and each filtering unit
comprises a plurality of despread filters, each said despread filter using one of
20 the plurality of despread vectors to despread a portion of the received CDMA
transmission encoded by the corresponding specific part of the spreading sequence.

28. The system of claim 26, wherein the interference cancellation unit
further calculates an error vector for each despread sequence and normalizes each
25 error vector, each normalized error vector used to calculate a corresponding one of the
at least one adapted despread sequence.

29. The system of claim 26, further comprising a channel estimation unit
for adapting the at least one despread sequence to account for radio channel
30 variations, wherein said adapting to account for radio channel variations is performed

- 24 -

according to an amplitude and a phase associated with portions of the received CDMA transmission.

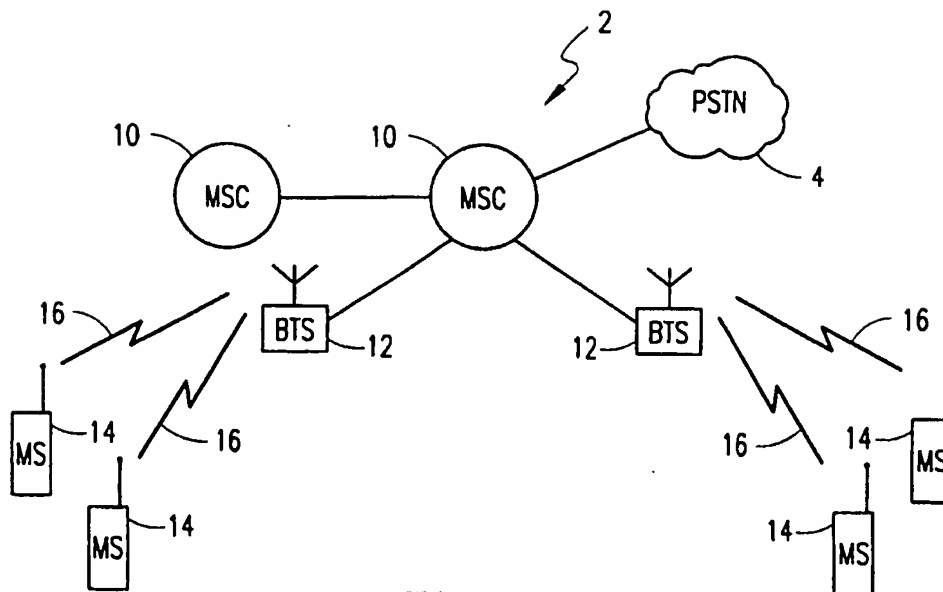


FIG. 1

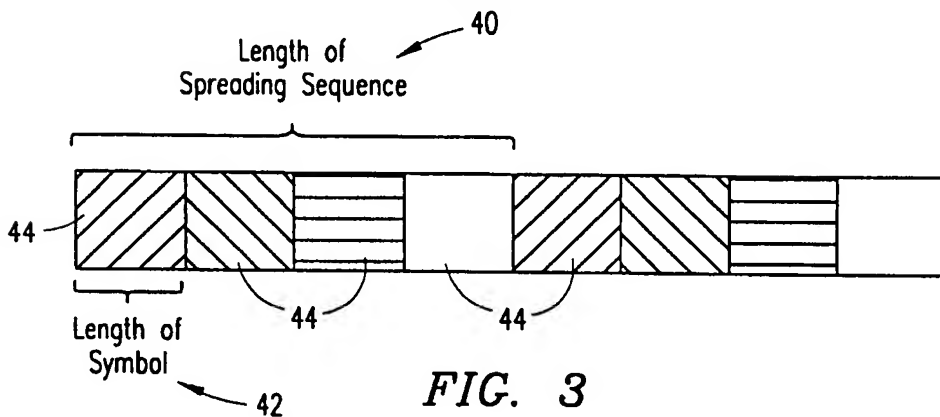


FIG. 3

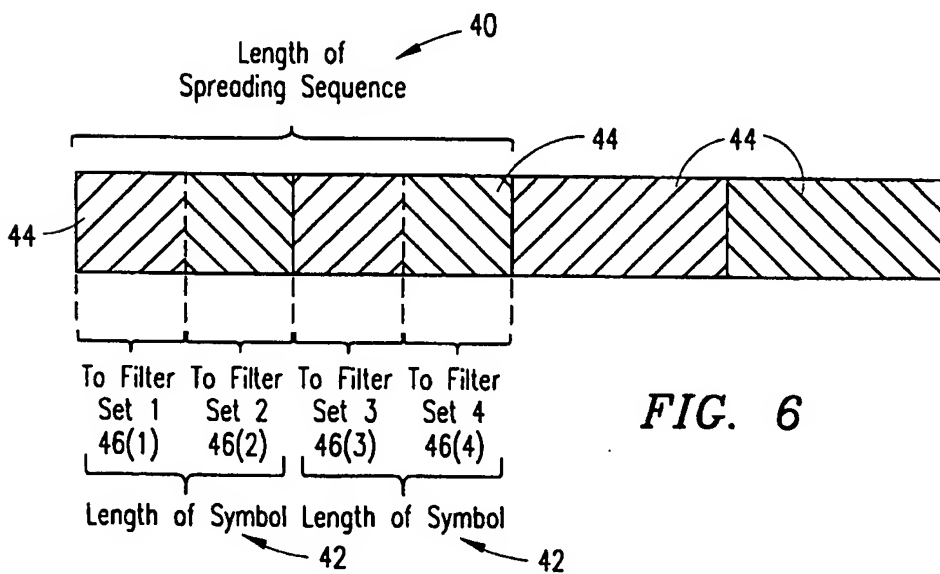


FIG. 6

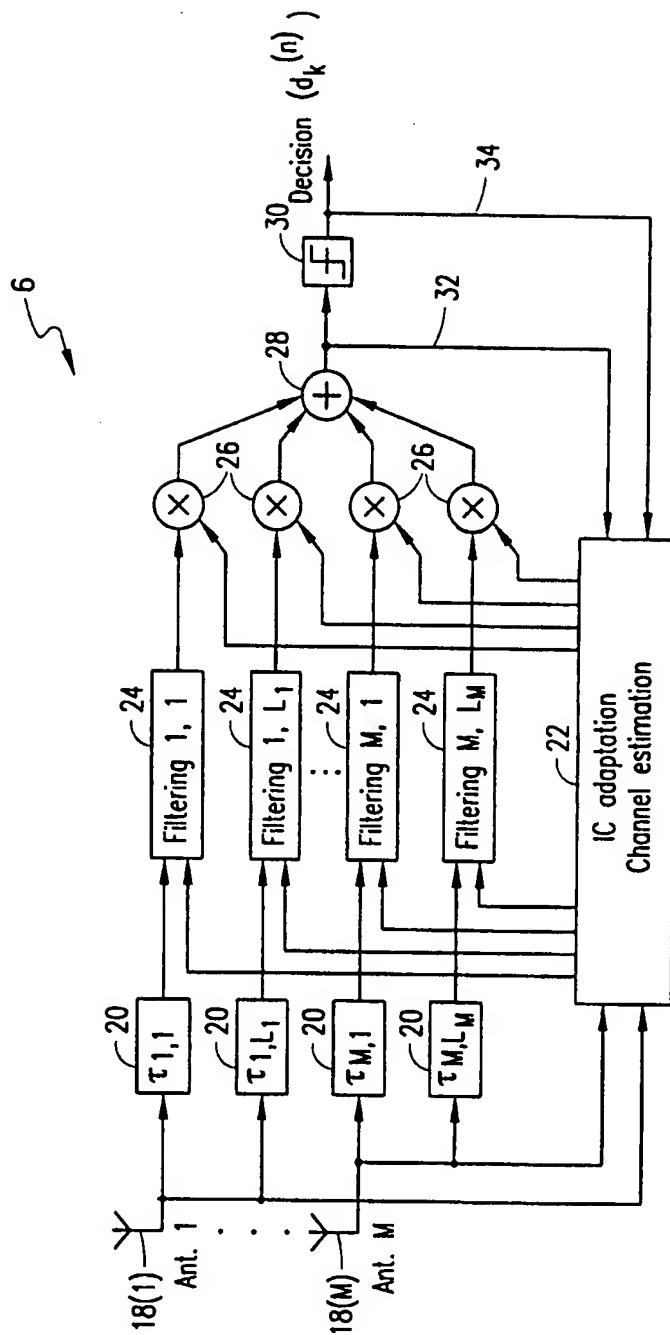


FIG. 2

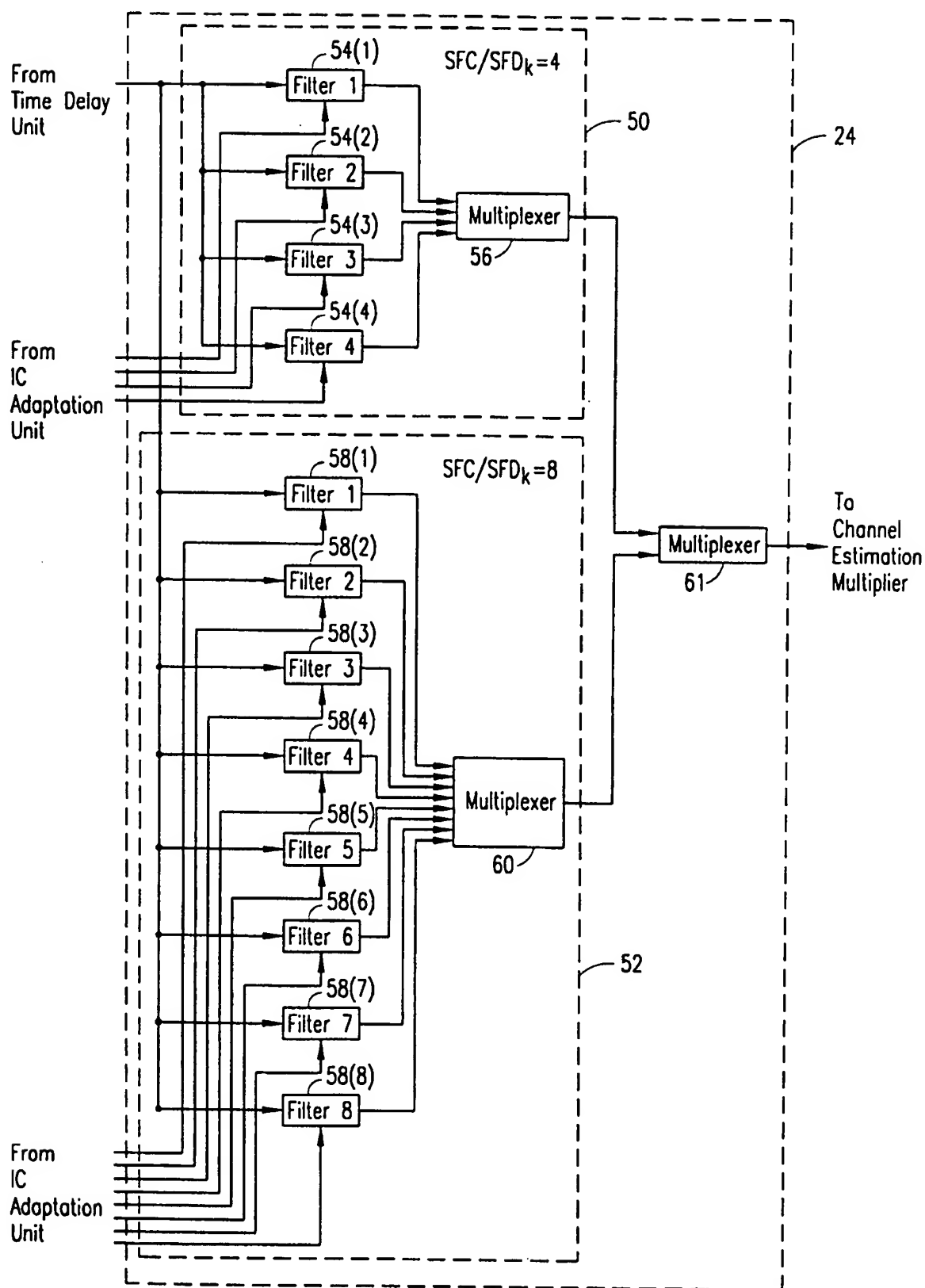


FIG. 4

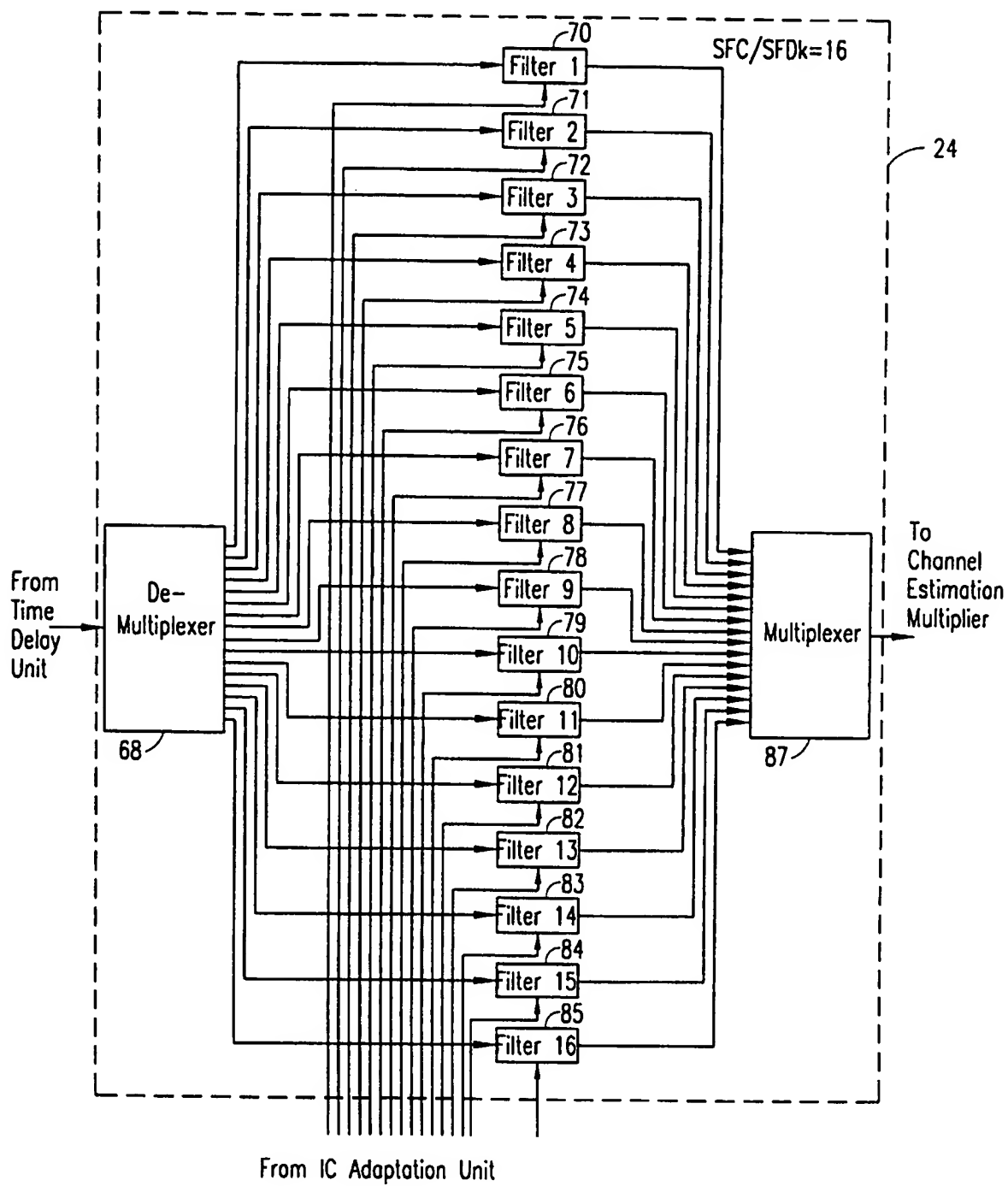


FIG. 5

INTERNATIONAL SEARCH REPORT

International Application No

PC1/SE 00/02067

A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 H04B1/707

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 H04B

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, INSPEC, WPI Data, PAJ

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	EP 0 936 749 A (EUROP AGENCE SPATIALE)	1,5,8
Y	18 August 1999 (1999-08-18)	
A	page 3, line 15 -page 4, line 8	14
	page 6, line 8 -page 7, line 1	2,6,9,
		15,20,26
	page 14, line 43 - line 52	
	page 15, line 41 -page 16, line 40;	
	figures 3,19	

	-/--	



Further documents are listed in the continuation of box C.



Patent family members are listed in annex.

* Special categories of cited documents :

- *A* document defining the general state of the art which is not considered to be of particular relevance
- *E* earlier document but published on or after the international filing date
- *L* document which may throw doubts on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- *O* document referring to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

- *T* later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
- *X* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- *Y* document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.
- *G* document member of the same patent family

Date of the actual completion of the international search

7 February 2001

Date of mailing of the international search report

15/02/2001

Name and mailing address of the ISA

European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 epo nl,
Fax: (+31-70) 340-3016

Authorized officer

Nilsson, M

INTERNATIONAL SEARCH REPORT

Inter "ional Application No

PC, /SE 00/02067

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	<p>VAN DER VEEN A -J: "Blind subspace-based uplink receiver algorithm for wideband CDMA"</p> <p>GATEWAY TO 21ST CENTURY COMMUNICATIONS VILLAGE. VTC 1999-FALL. IEEE VTS 50TH VEHICULAR TECHNOLOGY CONFERENCE (CAT. NO.99CH36324), GATEWAY TO 21ST CENTURY COMMUNICATIONS VILLAGE. VTC 1999-FALL. IEEE VTS 50TH VEHICULAR TECHNOLOGY CONFERENCE, AMSTERDAM,, pages 176-180 vol.1, XP002151990 1999, Piscataway, NJ, USA, IEEE, USA ISBN: 0-7803-5435-4 paragraphs '0002!', '0003!', '05.1!</p>	1,15,20, 26
Y	<p>HEIKKILA M J ET AL: "Interference suppression in CDMA downlink through adaptive channel equalization"</p> <p>GATEWAY TO 21ST CENTURY COMMUNICATIONS VILLAGE. VTC 1999-FALL. IEEE VTS 50TH VEHICULAR TECHNOLOGY CONFERENCE (CAT. NO.99CH36324), GATEWAY TO 21ST CENTURY COMMUNICATIONS VILLAGE. VTC 1999-FALL. IEEE VTS 50TH VEHICULAR TECHNOLOGY CONFERENCE, AMSTERDAM,, pages 978-982 vol.2, XP002151991 1999, Piscataway, NJ, USA, IEEE, USA ISBN: 0-7803-5435-4 paragraphs '0002!-'0004!</p>	14
A		20-23

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/SE 00/02067

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
EP 0936749 A	18-08-1999	FR 2774831 A	13-08-1999
		JP 11340952 A	10-12-1999
<hr/>			